

Industrial Process Control MDP 454

If you have a smart project, you can say "I'm an engineer" ??

Lecture 5

Staff boarder

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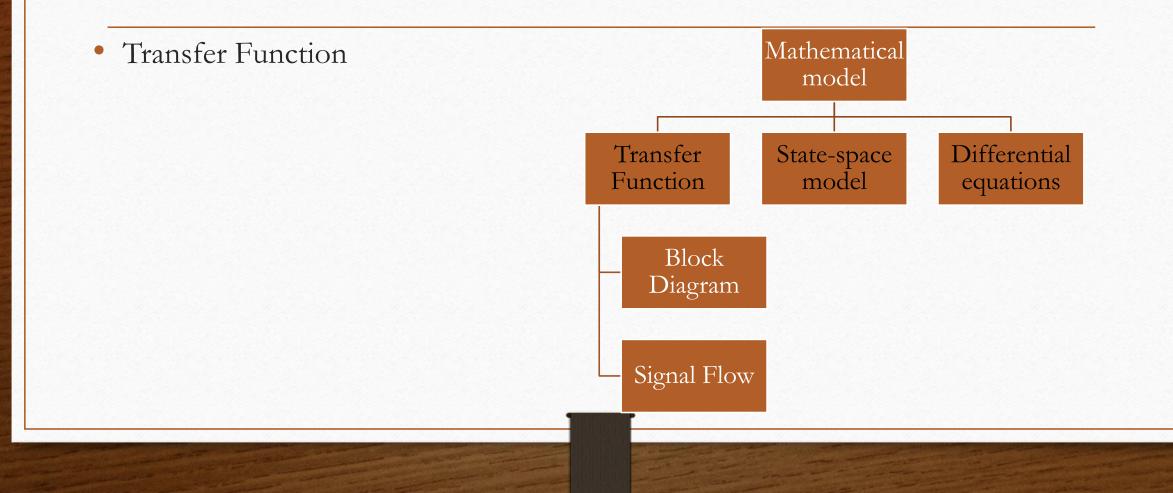
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Industrial Process Control MDP 454

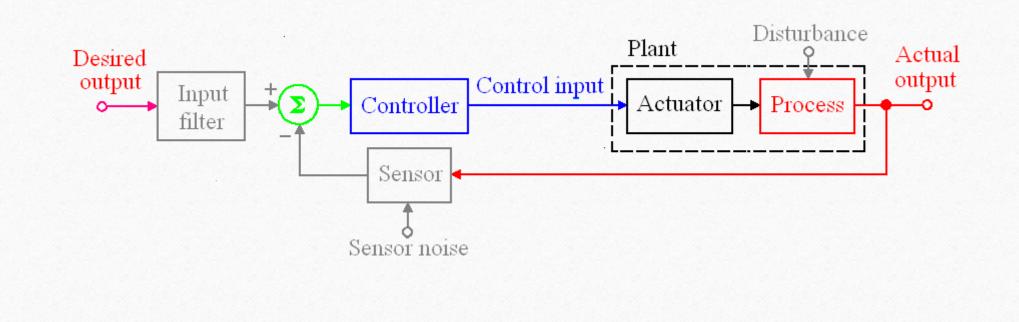
• Lecture aims:

- Understand the Block reduction techniques
- Identify the transfer function
- Be aware by modeling multiple technique

Mathematical Modeling

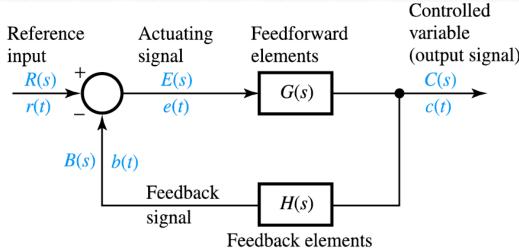


Component Block Diagram



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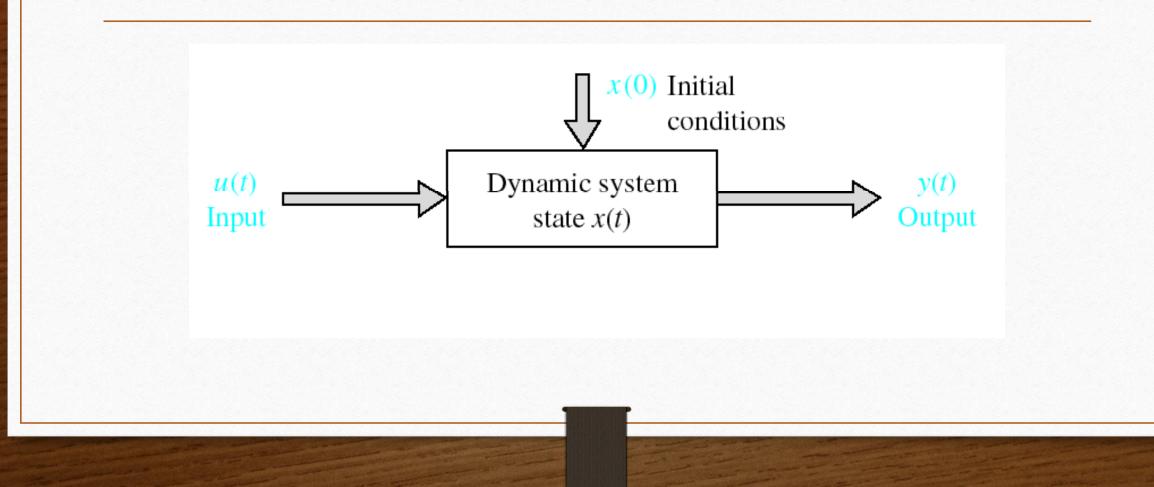
Component Block Diagram



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1	R(s)	Reference input
m ol)	C(s)	Output signal (controlled variable)
gnal)	B(s)	Feedback signal = $H(s)C(s)$
	E(s)	Actuating signal (error) = $[R(s) - B(s)]$
	G(s)	Forward path transfer function or
		open-loop transfer function = $C(s)/E(s)$
	M(s)	Closed-loop transfer function = $C(s)/R(s) = G(s)/[1 + G(s)H(s)]$
	H(s)	Feedback path transfer function
	G(s)H(s)	Loop gain
	E(s)	
	$\frac{1}{R(s)}$	= Error-response transfer function $\frac{1}{1 + G(s)H(s)}$

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The general form of a dynamic system



The general form of a dynamic system



State Space Equations

 $\mathbf{A} = \begin{bmatrix} a_{11} & a_{12} & \cdots & a_{1n} \\ a_{21} & a_{22} & \cdots & a_{2n} \\ \vdots & \vdots & & \vdots \end{bmatrix}$

 $\mathbf{B} = \begin{bmatrix} b_{11} & b_{12} & \cdots & b_{1m} \\ b_{21} & b_{22} & \cdots & b_{2m} \\ \vdots & \vdots & & \vdots \\ b_{2m} & b_{2m} & b_{2m} \end{bmatrix}$

 $\mathbf{D} = \begin{bmatrix} d_{11} & d_{12} & \cdots & d_{1m} \\ d_{21} & d_{22} & \cdots & d_{2m} \\ \vdots & \vdots & & \vdots \\ d_{21} & d_{22} & \cdots & d_{2m} \end{bmatrix}$

State equations is a description which relates the following $\dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t)$ four elements: input, system, state variables, and output $\mathbf{y}(t) = \mathbf{C}\mathbf{x}(t) + \mathbf{D}\mathbf{u}(t)$

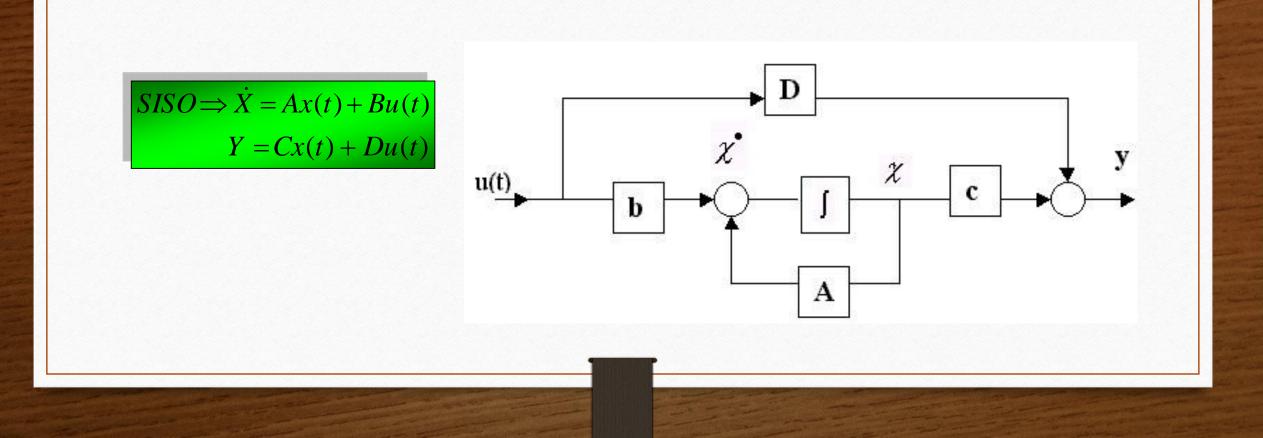
Matrix A has dimensions nxn and it is called the system matrix, having the general form

Matrix B has dimensions nxm and it is called the **input** matrix, having the general form $\mathbf{C} = \begin{bmatrix} c_{11} & c_{12} & \cdots & c_{1n} \\ c_{21} & c_{22} & \cdots & c_{2n} \\ \vdots & \vdots & & \vdots \end{bmatrix}$ Matrix C has dimensions pxn and it is called the

output matrix, having the general form

Matrix D has dimensions pxm and it is called the **feedforward** matrix, having the general form

State Space Equations



The general form of a dynamic system

The concept of a set of state variables that represent a dynamic system can be illustrated in terms of the spring-mass-damper system. A set of state variables sufficient to describe this system includes the position and the velocity of the mass.

We will define a set of state variables as (x1, x2), where

$$x_1(t) = y(t)$$
 and $x_2(t) = \frac{dy(t)}{dt}$. $\frac{dx_1}{dt} = x_2$

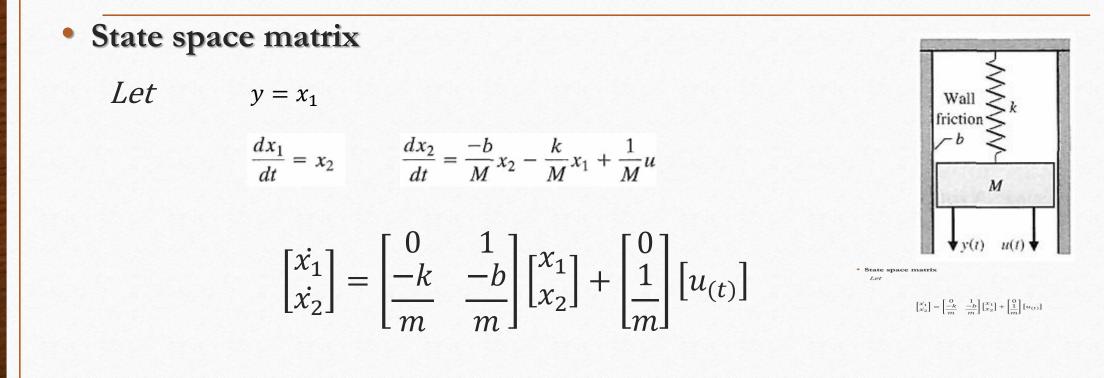
To write Equation of motion in terms of the state variables, we substitute the state variables as already defined and obtain $M\frac{d^2y}{dt^2} + b\frac{dy}{dt} + ky = u(t)$ de

$$M\frac{dx_2}{dt} + bx_2 + kx_1 = u(t)$$

Therefore, we can write the equations that describe the behavior of the spring-mass damper system as the set of two first-order differential equations

$$\frac{dx_2}{dt} = \frac{-b}{M}x_2 - \frac{k}{M}x_1 + \frac{1}{M}u$$

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RLC circuit example

• The state of this system can be described by a set of state variables (x1, x2), where x1 is the capacitor voltage vc(t) and x2 is the inductor current $i_L(t)$.

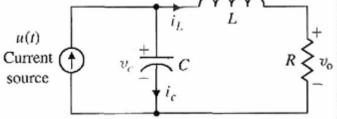
• Utilizing Kirchhoff's current law at the junction

OR

 $i_c = C \frac{dv_c}{dt} = +u(t) - i_L$

Kirchhoff's voltage law for the right-hand loop provides the equation describing the rate of change of inductor current as di_L

$$L\frac{di_L}{dt} = -Ri_L + v_c$$



The output of this system is represented

$$v_{\rm o} = Ri_L(t)$$

RLC circuit example

• rewrite Equations as a set of two first-order differential equations in terms of the state variables x1 and x2 as follows: $dx_1 = 1$, 1, 1, $dx_2 = 1$, R

$$\frac{dx_1}{dt} = -\frac{1}{C}x_2 + \frac{1}{C}u(t) \qquad \frac{dx_2}{dt} = +\frac{1}{L}x_1 - \frac{R}{L}x_2$$

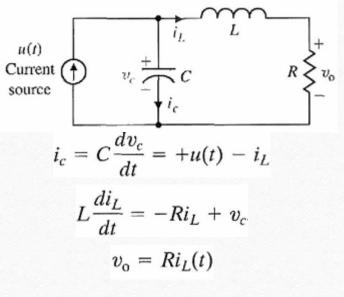
• The output signal is then

$$y_1(t) = v_0(t) = Rx_2$$

• obtain the state variable differential equation for the RLC

• and the output as

$$y = \begin{bmatrix} 0 & \frac{-1}{C} \\ \frac{1}{L} & \frac{-R}{L} \end{bmatrix} \mathbf{x} + \begin{bmatrix} \frac{1}{C} \\ 0 \end{bmatrix} u(t)$$



TRANSFER FUNCTION FROM THE STATE EQUATION

Obtain a transfer function G(s), Given the state variable equations. Recalling Equations :where v is the single output and u is the single input. $\dot{x} = Ax + Bu$ y = Cx + Du

The Laplace transforms of Equations sX(s) = AX(s) + BU(s)Y(s) = CX(s) + DU(s)

where B is an $n \ge 1$ matrix, since u is a single input, we obtain

$$(s\mathbf{I} - \mathbf{A})\mathbf{X}(s) = \mathbf{B}U(s)$$
$$\mathbf{X}(s) = \Phi(s)\mathbf{B}U(s)$$

• we obtain state transition Matrix

$$[s\mathbf{I} - \mathbf{A}]^{-1} = \mathbf{\Phi}(s)$$

TRANSFER FUNCTION FROM THE STATE EQUATION

• Transfer function
$$G(s)$$
: $G(s) = Y(s)/U(s)$ is

 $G(s) = \mathbf{C} \Phi(s) \mathbf{B} + \mathbf{D}$

• Let us determine the transfer function G(s) = Y(s)/U(s) for the *RLC* circuit, described by the differential equations

$$\dot{\mathbf{x}} = \begin{bmatrix} 0 & \frac{-1}{C} \\ \frac{1}{L} & \frac{-R}{L} \end{bmatrix} \mathbf{x} + \begin{bmatrix} \frac{1}{C} \\ 0 \end{bmatrix} u \qquad \qquad u(t) \\ Current \\ source \qquad \qquad v_c + \frac{i_L}{C} \\ \frac{1}{C} \\ \frac$$

TRANSFER FUNCTION FROM THE STATE EQUATION

Then we have $[s\mathbf{I} - \mathbf{A}] = \begin{vmatrix} s & \frac{z}{C} \\ \frac{-1}{L} & s + \frac{R}{L} \end{vmatrix}$ u(t)• Therefore, we obtain source $\Phi(s) = [s\mathbf{I} - \mathbf{A}]^{-1} = \frac{1}{\Delta(s)} \begin{bmatrix} \left(s + \frac{R}{L}\right) & \frac{-1}{C} \\ \frac{1}{r} & s \end{bmatrix} \Delta(s) = s^2 + \frac{R}{L}s + \frac{1}{LC}$ $G(s) = \begin{bmatrix} 0 & R \end{bmatrix} \begin{bmatrix} \frac{s + \frac{R}{L}}{\Delta(s)} & \frac{-1}{C\Delta(s)} \\ \frac{1}{2} & \frac{-1}{C\Delta(s)} \end{bmatrix} \begin{bmatrix} \frac{1}{C} \\ 0 \end{bmatrix} = \frac{R/(LC)}{\Delta(s)} = \frac{R/(LC)}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$ Then the transfer function is

• Transfer from time domain to frequency domain:

$$R_{1}i_{1}(t) + \frac{1}{C} \int_{0}^{t} i_{1}(t) dt - \frac{1}{C} \int_{0}^{t} i_{2}(t) dt = v(t)$$

$$\left[R_{1} + \frac{1}{Cs}\right]I_{1}(s) - \frac{1}{Cs}I_{2}(s) = V(s)$$

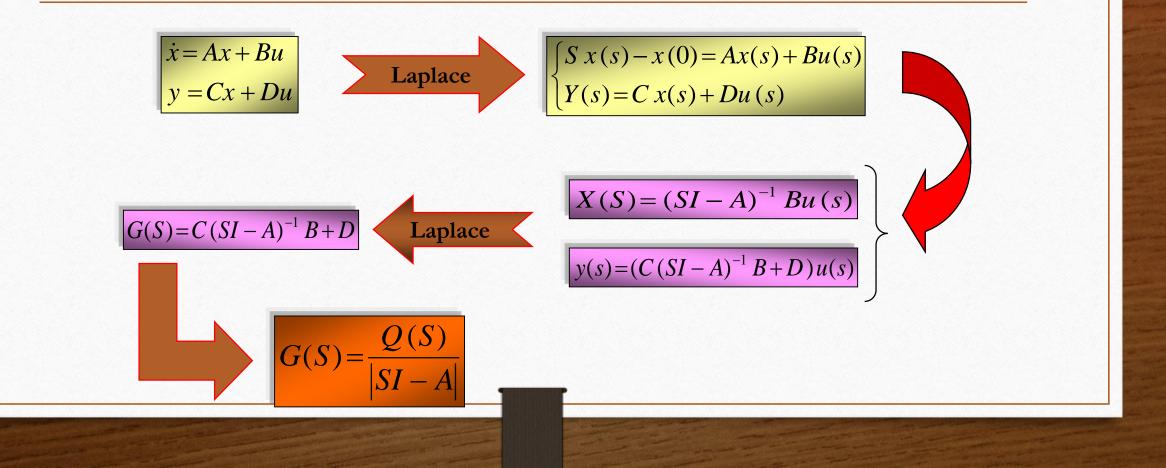
$$-\frac{1}{C} \int_{0}^{t} i_{1}(t) dt + R_{2}i_{2}(t) + L\frac{di_{2}}{dt} + \frac{1}{C} \int_{0}^{t} i_{2}(t) dt = 0$$

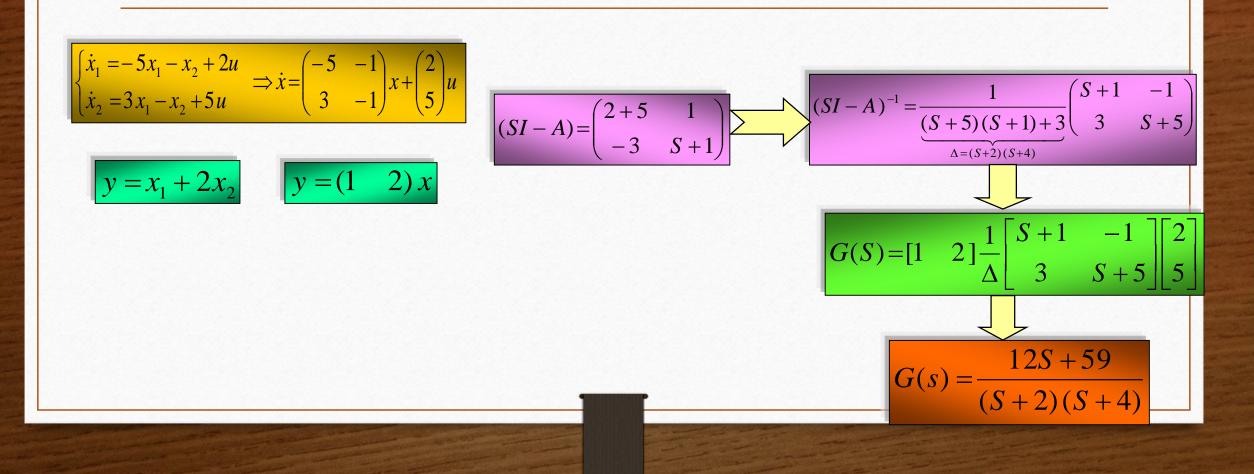
$$-\frac{1}{Cs}I_{1}(s) + \left[R_{2} + Ls + \frac{1}{Cs}\right]I_{2}(s) = 0$$
• Transfer function
$$\frac{I_{2}(s)}{V(s)} = \frac{Cs}{(R_{1}Cs + 1)(LCs^{2} + R_{2}Cs + 1) - 1} = \frac{1}{R_{1}LCs^{2} + (R_{1}R_{2}C + L)s + R_{1} + R_{2}}$$

$$\begin{cases} e(t) - R_1 i_1(t) - L_1 \frac{di_1}{dt} - V_C(t) = \phi \\ V_C(t) - L_2 \frac{di_2}{dt} - R_2 i_2 = \phi \\ i_c = i_1 - i_2 = C \frac{dv_c}{dt} \end{cases}$$

$$x = (i_1 \ i_2 \ v_c)^T$$

$$X^{\bullet} = \begin{pmatrix} -\frac{R_{1}}{L_{1}} & 0 & \frac{-1}{L_{1}} \\ 0 & \frac{-R_{2}}{L_{2}} & \frac{1}{L_{2}} \\ \frac{1}{C} & \frac{-1}{C} & 0 \end{pmatrix} \begin{pmatrix} x_{1} \\ x_{2} \\ x_{3} \end{pmatrix} + \begin{pmatrix} \frac{1}{L_{1}} \\ 0 \\ 0 \end{pmatrix} e(t)$$
$$0 = (t)$$





Model Examples

• Pulse Width Modulation (PWM)

